

RHYTHMICITY IN NEURON-LIKE LINES

BY

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THESIS

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INTRODUCTION

A great deal of effort has been made in the empirical investigation of the electrical properties of animal neurons and in the synthesis of neural analogs. In both cases studies have been

largely confined to the case of the propagation of a single pulse, lasting only a few milliseconds. Such behavior is often referred to as 'all or none' response. Essentially, it represents one

'bit' of information. Only by repetitive behavior is the information capacity of the line increased. It is the intention of

this paper to delve into the possibilities of recurrent as opposed to 'all or none' behavior of these lines when d.c. voltage inputs

are applied.

To undertake this study, two well established models will be

utilized. The first, which is based solely on laboratory findings

is the Hodgkin-Huxley<sup>1</sup> model of the giant squid axon. The second

is the Neustor<sup>2</sup> model which is itself based upon a simplification

of the Hodgkin-Huxley model. Both models are reasonably consistent

with empirical data. Both will be employed to investigate differ-

ent modes by which recurrent or rhythmic behavior can be achieved.

In this endeavor we shall approach our problem from an engi-

neering standpoint. We will examine possible modes of repetitive

discharge as purely electrical phenomena, and appeal to our know-

ledge of electrical circuit theory.

<sup>1</sup> A. L. Hodgkin, A. F. Huxley, "A quantitative description of membrane current and its application to conduction and excitation of nerve," Journal of Physiology, vol. 117, pp. 500-540; August, 1952.

J. Nagumo, S. Arimoto, S. Yoshizawa, "An active pulse transmission line simulating nerve axon," Proceedings of the IRE, vol. 50, no. 10, pp. 2061-2070; October 1962.

Experimental data with animal neurons known as receptor cells indicate that these lines respond to d.c. input voltages (referred to as generator potentials) with a train of spike-like pulses. It is generally acknowledged that the pulse frequency varies directly with the input voltage, (though frequency may gradually decline or adapt). Yet histological examination shows that the receptor cell has characteristics similar to all nerves though its response is not the same. It is logical that we begin our investigation of rhythmicity by first looking at neuron-like lines, both physio-logical and electrical in nature.

## THE NEURON-LIKE LINE

There are of course other models that we might choose to

elaborate on. Let us therefore define descriptively the properties of the neuron-like line in general. Here we might select a variety of characteristics associated with some animal neuron. Thus the following definition is somewhat arbitrary. The neuron-like lines to be discussed will have these properties:

1. Distributed parameters; which are uniform with respect to distance.

2. Bidirectional characteristics (except for end point discontinuities).

3. 'All or none' action potential response.

Some explanation of the third property is in order. In a few words, such a line, when excited at any point along its length, will respond in one of two (all or none) manners, depending upon the magnitude and duration of the excitation:

a. (none) - Below the so-called 'threshold' level of excitation.

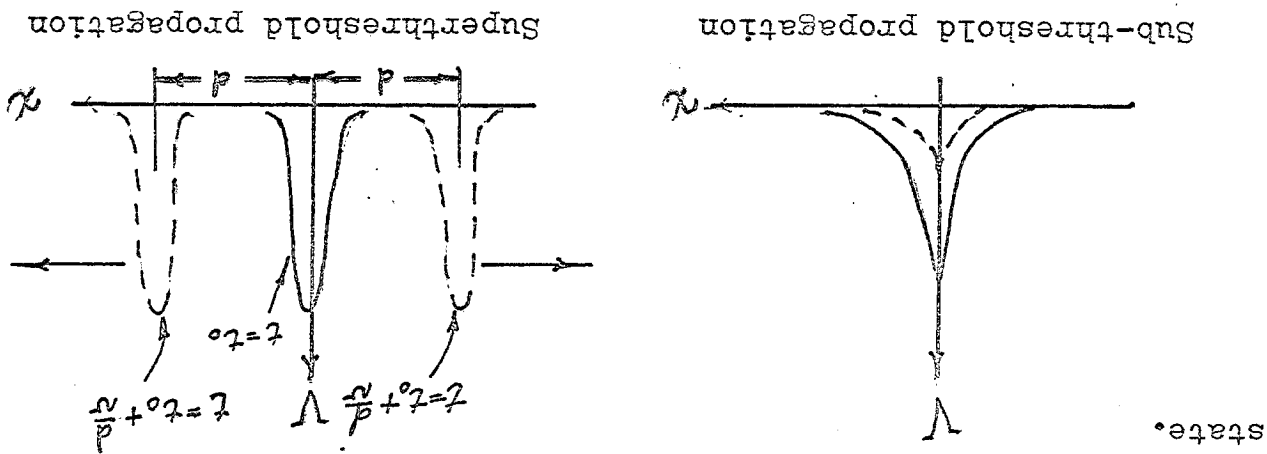
b. (all) - Above the threshold level of excitation, an

tion.

'action potential' wave with a shape which is independent of the level of excitation, propagates away from the point of excitation in both directions at a uniform velocity.

The animal neuron axon and certain of its analogs such as the Neuristor have all of the above properties. Such lines behave linearly and passively with high attenuation at sub-threshold levels, and nonlinearly and actively with zero attenuation at super-threshold levels. These lines 'fire' electrically much the same as fuses when heated above some ignition point, but they are

recoverable after a brief 'refractory' period to their original state.



$V$  = line voltage

$x$  = line distance

$t_0$  = time at which excitation is applied

$v$  = velocity of propagation

Fig. 1. Propagation of the action potential.

THE LINEAR SUB-THRESHOLD LINE

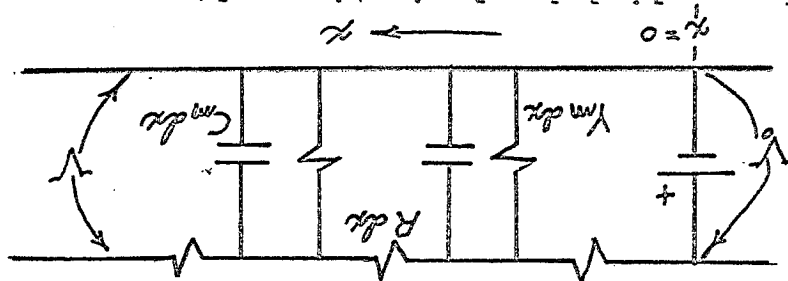
Electrotonic Spread

Now that we have some notion of the neuron-like line, let

us proceed to study analytically a typical line. Below threshold

depolarization (line voltage) the line behaves passively as a

uniform cable having a simple linear model.



$V^0$  = applied depolarization voltage

$C_m^m$  = shunt capacitance/unit length

$R$  = series resistance/unit length

$Y_m^m$  = shunt conductance/unit length

Fig. 2. Linear, sub-threshold neuron circuit model.

In the steady-state we may neglect the effects of distributed

shunt capacitance so that an applied voltage,  $V^0$  at  $x=0$  decays

exponentially according to the relation,

$$V = V^0 \exp + x(RY_m)^{-1/2} \quad x < 0 \quad (1)$$

$$V = V^0 \quad x = 0 \quad (2)$$

$$V = V^0 \exp - x(RY_m)^{-1/2} \quad x > 0 \quad (3)$$

This exponential variation is referred to in electrophysiology

as electrotonic spread and we shall investigate it in greater detail

later.



THE NONLINEAR SUPER-THRESHOLD LINE

Hodgkin-Huxley and Neuron Models

Now let us consider what occurs when the applied voltage  $V^0$  exceeds the threshold level,  $V^0 > V^{th}$ . Strictly speaking, corresponding to  $V^0(t)$  we encounter  $V^{th}(t)$ ; that is, the threshold is a function of the way in which  $V^0$  varies with time (accommodation).

In this case our neuron line becomes active; it 'fires'.

our linear model breaks down, and an action potential propagates in both directions down the line. This break-down has been studied extensively by Hodgkin and Huxley (and others). Their nonlinear model presents analytical difficulties, but it is subject to a number of reasonable simplifications.

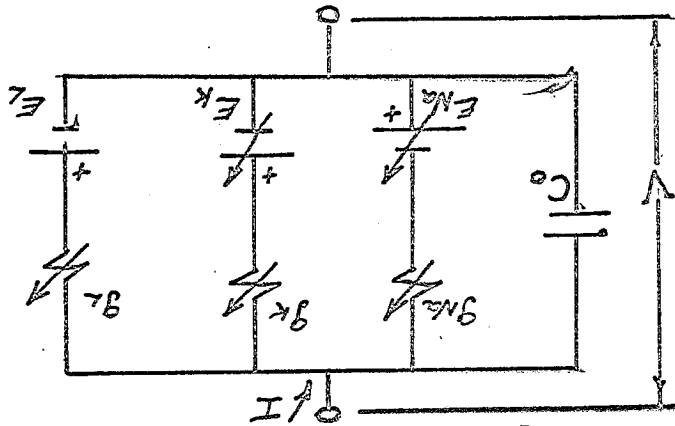


Fig. 3. Hodgkin-Huxley neuron circuit model.

The above model represents many years of experimental work

with the giant squid axon. The equations which describe this

model are given below. Although little use will be made of them,

they are presented to indicate the complexities inherent in our

study of neuron behavior. Total membrane current is given by

$$I = C_0 \dot{V} + \bar{g}_{Na} m^2 h (V - E_{Na}) + \bar{g}_K n^4 (V - E_K) + \bar{g}_L (V - E_L) \quad (4)$$

$$m^0 + (\infty_m + \beta_m) m = \infty_m \quad (5)$$

4 A.L. Hodgkin, A.L. Huxley, op. cit.

(14)  $1 > b > 0, c^2 > b, 1 > a > 1 - \frac{2}{3}b$

where a, b, and c are constants satisfying the relations.

(13)  $C_0^w + b_w = a - \mu$

(12)  $f = \frac{c}{\tau} \mu - (\mu - w - \mu^2/3)$

the following mathematical model (Bonhoeffer-van der Pol, BVP). (e.g. tunnel diode behavior). The Neurtistor is often based upon puted elements exhibiting R, L, C and nonlinear characteristics

Neurtistor. Theoretically, such a line is constructed of distri- Perhaps the best electrical analog to the above model is the useful later at least qualitatively.

possible modes of rhythmic behavior. This model will however be is obviously laborious and it gives us essentially no insight into of Hartree) to these partial differential equations. The method equation solution and applying successive approximations (method gation of an action potential is obtained by assuming a wave From these 'space clamp' equations (V=constant) the propa-

The above was derived entirely from empirical results.

(11)  $\beta_m = .125 \exp(\frac{v}{10})$  (where  $\beta_0 = \frac{d\beta}{dv} \frac{dv}{dt}$  at  $t = t_{imp}$ )

(10)  $\alpha_n = .01(V+10) [\exp(\frac{V+10}{10}) - 1]^{-1}$

(9)  $\beta_h = (\exp[\frac{V+30}{10}] + 1)^{-1}$

(8)  $\beta_m = \tau \exp(\frac{V}{10}), \alpha_h = .07 \exp \frac{20}{V}$

(7)  $\alpha_m = .1(V+25) [\exp(\frac{V+25}{10}) - 1]^{-1}$

(6)  $\dot{y} + (\alpha_h + \beta_h)y = \alpha_h \dot{y} + (\alpha_n + \beta_n)n = \alpha_n$

and the variables  $u, w$  and  $j$  in the above correspond to the pair of variables  $(V, m)$ ,  $(h, n)$  and  $I$  in the Hodgkin-Huxley equations. An electronic simulator of the BVP model is shown below.

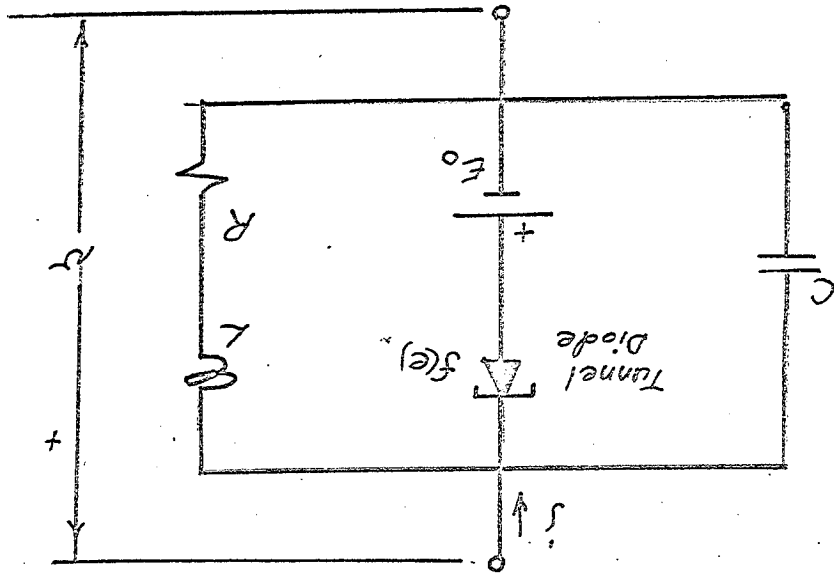


Fig. 4. One version of the tunnel diode discrete element neuristor.

6 J. Nagumo, S. Arimoto, S. Yoshizawa, op. cit. p. 2063.

Now that we have acquired some familiarity with the basic

properties and structure of neuron-like lines, we can begin probing into their rhythmic characteristics. Our dictionary defines

rhythm as: "Movement characterized by regular measured or harmonic recurrence of stress, beat, sound, accent, or motion; as the rhythm

of the pulse." More specifically we shall confine ourselves to

rhythmic response under input voltage control. It is by the

process of voltage controlled modulation of pulse frequency that

quantitative information is transmitted along the animal neuron.

This voltage controlled form of oscillation must be distinguished

from fixed frequency oscillation such as one finds in animal neurons

after the removal of calcium (decalcification). In the later case,

oscillations are independent of any input voltage and the system is

said to be unstable.

RHYTHMICITY AS A STABILITY PROBLEM

Ljapunov's Second Method

It is only natural to consider rhythmicity in terms of stability.

For-

Rhythmicity implies instability

The type of oscillations we are considering actually represent

conditional instability. That is, instability which is dependent

upon input voltage. If the input voltage is removed, no oscillations

occur, but if an input above threshold is applied, then pulsations

are present and their frequency increases with input voltage.

To test whether our system is capable of rhythmic response,

we can appeal to stability criteria. Is the system stable? If the

system is stable, it must have a Lyapunov function  $L(\bar{s})$  where  $\bar{s}$  denotes the state vector. Thus in both Hodgkin-Huxley and neuronistor equations we can seek out a scalar  $L(\bar{s})$  such that:

$$L(\bar{s}) > 0, \quad L'(\bar{s}) < 0, \quad \text{when } \bar{s} \neq \bar{s}_e, \quad (15)$$

where  $\bar{s}_e$  is the equilibrium state and

$$L(\bar{s}_e) = L'(\bar{s}_e) = 0 \quad (16)$$

If we are able to find  $L(\bar{s})$ , then unending rhythmicity is not possible. This does not exclude rhythmicity of a transient nature (i.e. adapted response). If we are unable to find an  $L(\bar{s})$ , this does not of course imply that one does not exist and that the system is unstable. To establish instability, we require a new scalar function, call it  $L(\bar{s})$  such that

$$L(\bar{s}) > 0, \quad L'(\bar{s}) > 0, \quad \text{when } \bar{s} \neq \bar{s}_e \quad (17)$$

$$\text{and } L(\bar{s}_e) = L'(\bar{s}_e) = 0 \quad (18)$$

We are lead immediately to Lyapunov's Second Method here,

because of its complete generality in dealing with nonlinear systems. The difficulty lies in its application. Also, assuming that some

$L(\bar{s})$  exists, we are still unable to say

$$L(\bar{s}) \implies \text{instability} \implies \text{rhythmicity}$$

(instability does not imply rhythmicity)

It is necessary to examine our models more closely for

rhythmic tendencies.

7 R.E. Kalman, J. E. Bertram, "Control system analysis and design via the 'second method of Lyapunov,'" Journal of Basic Engineering, Transactions of the ASME, pp. 371-400; June 1960.

As we can see, knowing that a line is unstable does not in itself identify the type of instability and the mechanism by which recurrent response is possible. There are, however, several modes of oscillation that we might consider. To motivate later development, the following are suggested.

Take the relaxation oscillator for example. In essence a relaxation oscillator is a nonsinusoidal waveform generator. A

typical relaxation device involves the repeated charge and discharge of an energy storing component, such as an inductor or capacitor through a nonlinear electronic switch such as a tunnel diode. The simplest circuit that we might imagine is a series R-C configuration with the capacitance shunted by some device which discharges the capacitance at some critical voltage.

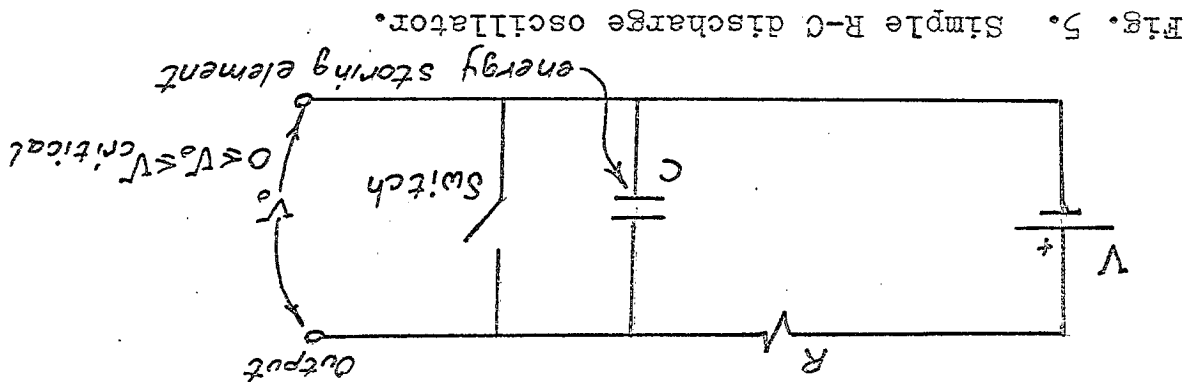


Fig. 5. Simple R-C discharge oscillator.

We note that the output frequency of this circuit will increase with  $V$ , as  $C$  charges to the critical voltage more rapidly with increasing  $V$ .

Neuron-like lines do contain series  $R$ , shunt  $C$  and nonlinear

'switching' properties, but these are all distributed. It is re-

quired that we elaborate further before accepting oscillations of the

relaxation type in neuron-like lines.

Yet another form of oscillation that we might associate with any distributed line involves line resonance. Consider the situation when a line is shorted at both ends. If it is lossless, then we know immediately that it is resonant at odd half-wave lengths. Consider next the case of a neuron-like line which is terminated in other than its characteristic impedance. (We have not defined the correct termination for such a nonlinear line but let us assume for the moment that it exists). Assume also, that at some point in the line behind its termination, another reflection is possible. Then we might again say that our line is resonant. The frequency is given by

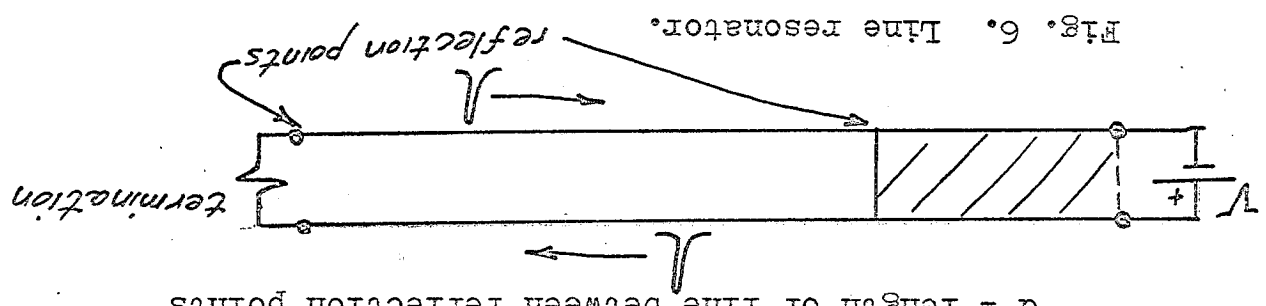
$$f = v/2d$$

(19)

where  $f$  = pulse frequency

$v$  = pulse propagation velocity, assumed constant

$d$  = length of line between reflection points



Also assume that the left-end reflection point is under voltage control. Nothing has been said about how this resonant line condition can be achieved in neuron-like lines. This shall be discussed at some length later.

Perhaps the most obvious oscillator is that which employs direct feedback. No further elaboration on this mode is necessary. However, it is difficult to see how feedback loops can develop in a distributed line with uniform characteristics. Secondly, how can the frequency of a closed loop oscillator be under voltage control?

Here, let us assume that our line exhibits continuous feedback.

Allow external uni-directional feedback from an endpoint discon-

tinuity into the line along its length. Also, let there be re-

generation (looping back) at some point on the line which is under

voltage control.

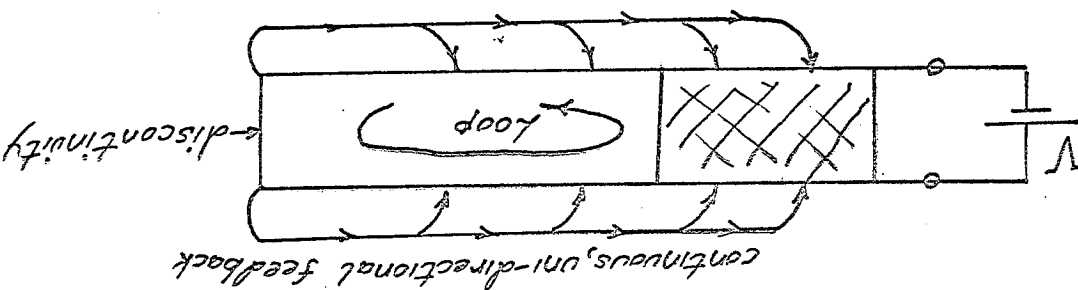


Fig. 7. Continuous feedback oscillator.

Admittedly, the concept of continuous feedback is difficult to accept, but we shall, however, not reject it as a possibility.



## CIRCUIT APPROXIMATIONS

At this stage, it is necessary to turn to the neuron-like models themselves for true insight into the mechanisms by which some of the above modes of oscillation are realizable. To be completely rigorous we should appeal directly to the sets of differential equations which describe these models. Such a procedure is, however, fraught with difficulties. In both cases analytical solutions are generally unattainable so that we must rely entirely upon numerical techniques and computer solutions. Any manipulation of these equations will only compound our problem.

Secondly, it is not our purpose here to obtain numerical results of limited applicability. Thus we must obtain algebraic results without analytical solutions.

To solve this dilemma, a number of simplifying assumptions will be made. This procedure will be followed entirely out of necessity. To test the validity of these assumptions we must avail ourselves of quantitative information which we shall not deal with here in this paper.

There are of course many simplifications of the neuron-like

line that we might make without stretching the imagination too far.

Let us first investigate a reasonable approximation of the Hodgkin-

Huxley model.

This model represents one section of the neuron line which is

space clamped at a constant voltage. An incremental extrapolation

should include series resistance as well as shunt conductance and

susceptance.

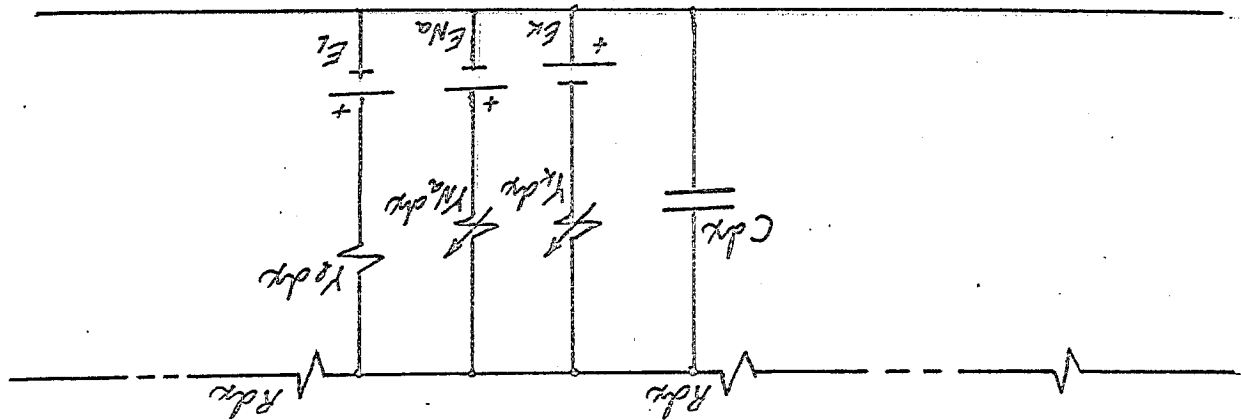


Fig. 8. Incremental extension of Hodgkin-Huxley circuit model.

This infinitesimal circuit reduction is achieved by spatially

chopping the Hodgkin-Huxley circuit down to a width  $dx$  and then

coupling it to similar sections with series resistance  $Rdx$ .

It should be noted that the shunt conductances have fixed

biasing voltages (equilibrium potentials,  $E_{Na}$ ,  $E_K$ ,  $E_L$ ). This was

done because each of the conductances was isolated experimentally

and it was discovered that the equilibrium potentials were essen-

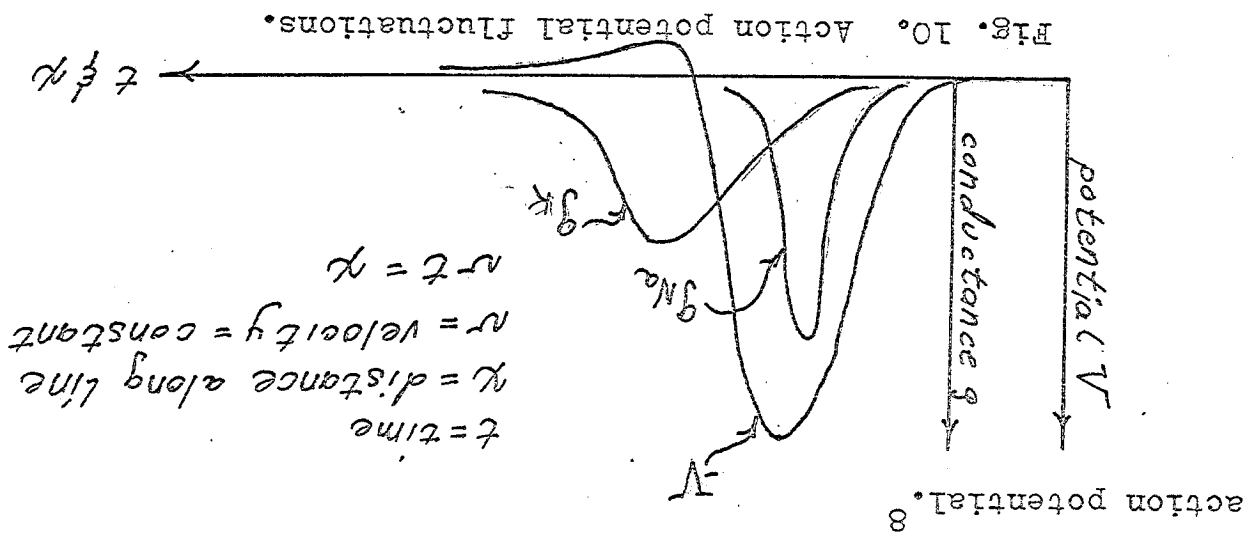
tially independent of applied voltage.

For convenience we may combine all the conductance branches

into one equivalent  $Y$  and an equivalent series voltage  $E$ . It must

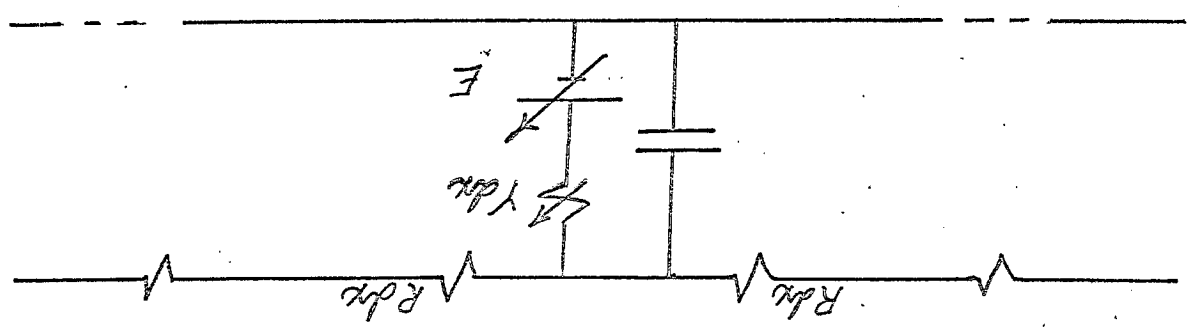
be noted that both  $Y$  and  $E$  are nonlinear. In the original model

The above is not quantitative, but it does clearly suggest that shunt conductance increases sharply with  $V$ . It is reasonable to assume that series resistance  $R$  is linear since the longitudinal voltage component is normally small. Although the transverse voltage  $V$  may be large, on the basis of experimental evidence,  $C$  is relative-



is suggested by the sketch below for the case of the propagated It must be remembered that  $V$  is a nonlinear function of  $V(x)$ . This a voltage  $V_0$  which is above threshold, over some region on the left. Now that we have simplified the model somewhat, let us apply

FIG. 9. Reduction of Hodgkin-Huxley incremental model.



one source, requires that it too be nonlinear and bipolar.  $V_K$  and  $V_{Na}$  vary sufficiently. To achieve the same response with reach zero and reverse (which it does during activation) when  $E_K$  and  $E_{Na}$  are opposite in sign, thus allowing the line voltage to

It is most important that we take note of the fact that both R and C parameters are constant, independent of the line voltage V. For small perturbations about some equilibrium point we may linearize the line parameters, gradually tapering their characteristics with respect to distance to conform to the line voltage.

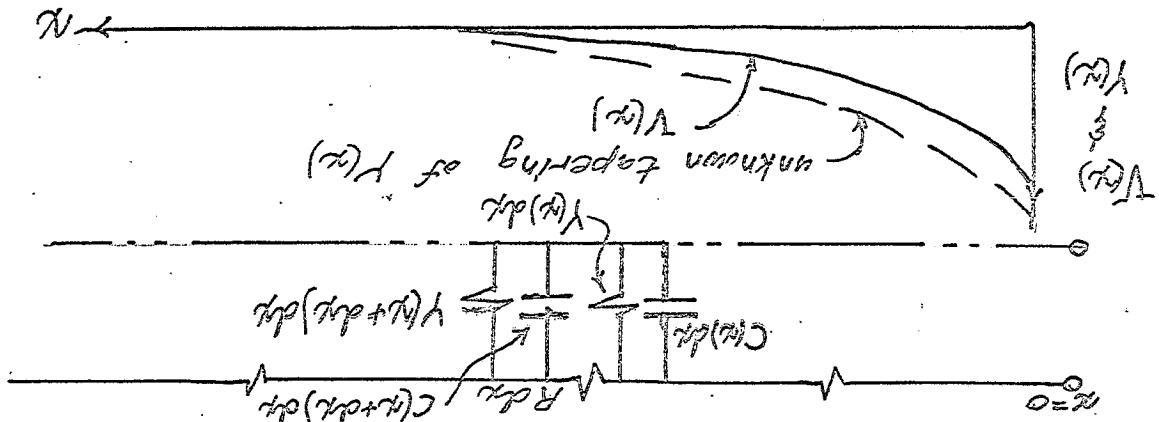


Fig. 11. Line parameter tapering.

The source E is removed in this model as we are assuming small perturbations. From left to right we expect to find constant R and C and decreasing Y. Thus we might divide the line artificially into three segments:

1. where Y is large relative to R,  $Y \gg R$  followed by a segment
2. where R is larger than Y,  $R > Y$  followed by a segment
3. where Y is negligible and shunt capacitance is important

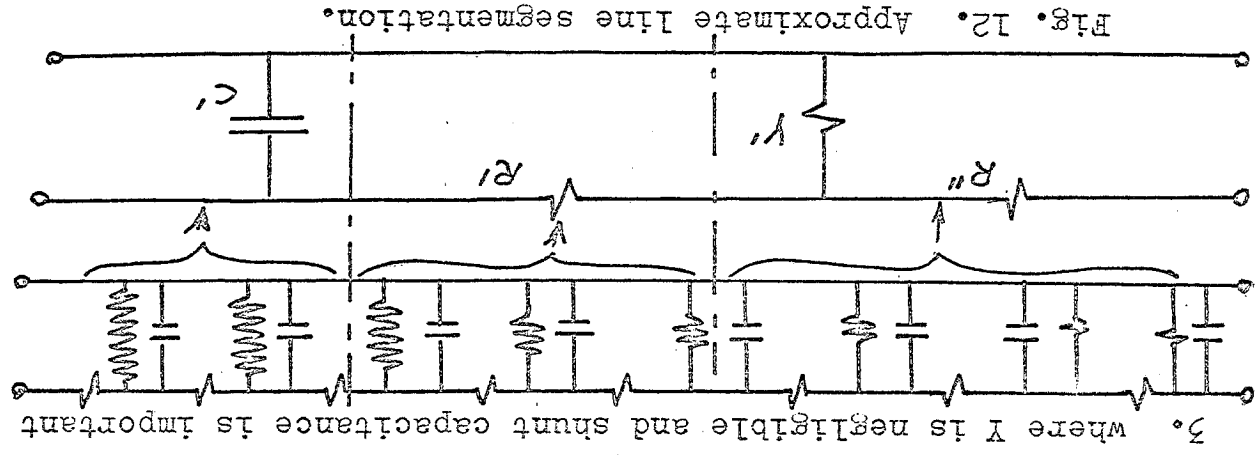


Fig. 12. Approximate line segmentation.

NOTE: The third section effective capacitance  $C'$  should approach

a limit as  $x \rightarrow \infty$  due to series resistance  $R$ .

Thus we have in a very crude fashion developed an R-C approx-

imation of an above-threshold line. We do not possess sufficient

evidence to justify this approximation. If we do accept this

approximation, it becomes easy to see how a relaxation oscillator

develops out of the model. It is only necessary to have  $C'$  alter-

nately charge and discharge to have recurrent behavior. This will

occur during the propagation and recovery phases of the action

potential.

The concept here is that of a super-threshold line on the verge

of firing, which is triggered by an R-C discharge.

It is interesting to note the properties of the R-C discharge

circuit.

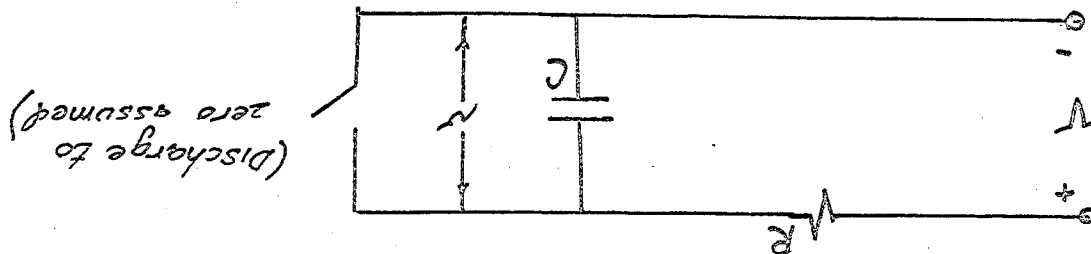


Fig. 15. R-C discharge oscillator.

$C$  is discharged when  $v = V_f$

For  $v < V_f$  we have

$$v = V_f (1 - e^{-t/RC})$$

(20)

As an infinite series the exponential term is given by

$$e^{-t/RC} = 1 - t/RC + t^2/2!R^2C^2 - t^3/3!R^3C^3 + \dots$$

(21)

which is approximated by

$$e^{-t/RC} \approx 1 - t/RC$$

(22)

cell.

This is the relationship that one commonly finds experimentally between generator potential  $V$  and pulse frequency  $f$  in the receptor

$$(25) \quad f = \text{frequency} \approx V \left( \frac{1}{VfRC} \right)$$

As the frequency varies directly with the applied voltage:

$$(24) \quad \tau = \frac{V}{VfRC} = \text{period of pulsation}$$

$$\text{when } V = Vf$$

$$(23) \quad V \approx \left( \frac{V}{RC} \right) \tau$$

$$\text{when } \tau/RC < 1$$

Let us take another look at the implications of our R-C

discharge approximation model.

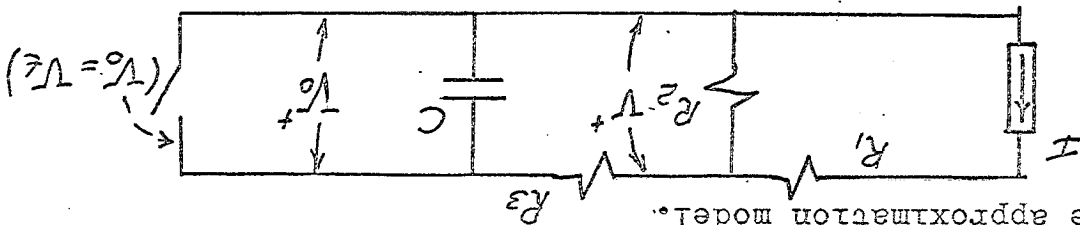


Fig. 14. R-C discharge oscillator with current input.

Assuming  $R_2 < R_3$  we have immediately

$$(26) \quad V \approx R_2 I = \text{constant}$$

and the output voltage is

$$(27) \quad V_0 = V (1 - e^{-t/R_2})$$

$$(28) \quad V = \frac{V_0 (1 - e^{-t/R_2})}{V_0}$$

and at the time of discharge  $V_0 = V_2$

$$(29) \quad V = \frac{V_2 (1 - e^{-t/R_2})}{V_2} = IR_2$$

$$(30) \quad I = \frac{V_2}{R_2} [1 - e^{-t/R_2}]$$

We have arrived at a relation between a constant excitatory

current and the time required to 'fire' the device. But this is

completely analogous to the empirical expression for nerve stimula-

tion which is

$$(31) \quad I = I_{th} [1 - e^{-t/k}]^{-1}$$

9 Bard, et al, op. cit., p. 925.

"The value of  $K$  depends on various parameters of the fiber. Some of these are membrane capacity, the length constant and the critical level of depolarization of the membrane required for excitation."<sup>9</sup>



APPROXIMATION 2 LINE RESONANCE

Concept of Heightened Excitability

As a second approach let us continue with a study of actual line resonance. In this case it is worthwhile to look into the neuronistor line. With this line we can most readily gain some

insight into wave propagation and repeated reflection accompanying line resonance. To eliminate difficulties inherent in any study of a nonlinear distributed line, we shall begin with an approximation of the neuronistor consisting of lumped parameters and relays. Basically, we may construct such a line by connecting a string of monostable multivibrators, each capable of firing the succeeding one with a finite delay. 10

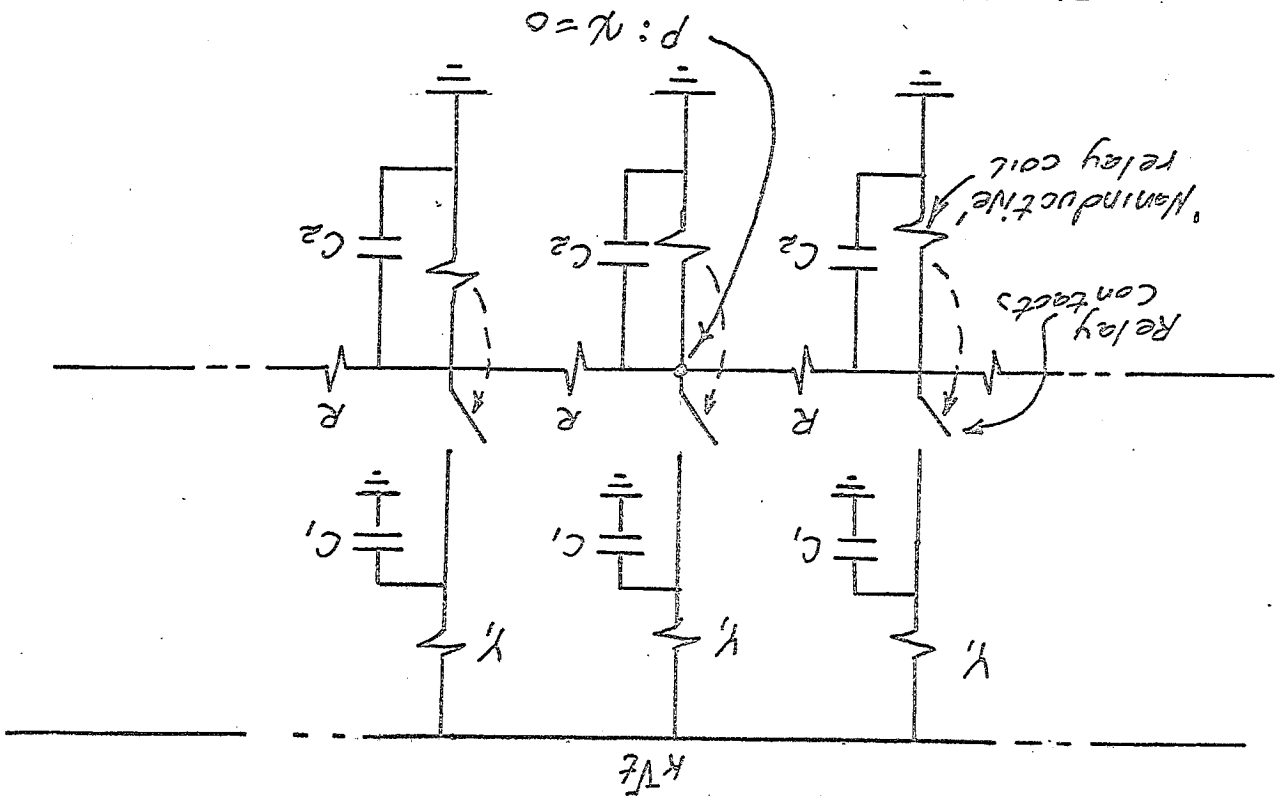


Fig. 15. Relay approximation of the Neuronistor

Designating  $V$ , applied at point  $P$  as the input, let us consider

$R, C$ , relay line.

model. To see how this is possible, let us return to the simple introduction of some form of retarded feedback into our original voltage control. The answer is yes, if we are willing to accept the the Neuronistor to achieve repetitive discharges of pulses under input Now we might ask ourselves if there is some way of modifying

plified Hodgkin-Huxley neuron model.

been experimentally synthesized which conforms closely to a sim- fact a neuronistor model using lumped  $R, C, L$  and tunnel diodes has is in accord with current concepts of the physiological neuron. In the line without attenuation and at some finite velocity. All this When the line is made active, an action potential traverses lumped parameter network to its transmission line extension).

in practice, we must realize the difficulty involved in reducing a at least there is no reason why it can not be distributed, (Although of the neuronistor line. It is uniform, nonlinear and conceptually At this point it would be profitable to review the properties propagation time delay.

disturbance will propagate. Note  $C_2$  is introduced to achieve a relays on each side of point  $P$  will have a chance to fire and the at  $P$  (for  $C_2 \ll C_1$ ). For  $k$  sufficiently greater than 1, succeeding and  $kV_t$ , the voltage to which  $C_1$  is charged is suddenly impressed along the line. However, when  $V(0) > V_t$  the relay contacts close  $V > V_t$ , the relay actuating voltage,  $V$  will decay electronically Suppose a voltage  $V$  is applied at point  $P$ . As long as

characteristics will for the present serve our purpose.

This model, although a gross simplification of neuron-like

the response of the line for  $x \gg 0$  for sub- and super-threshold  $V$ , ( $V < V_t$  and  $V > V_t$ ). From earlier discussion, we have that  $V$  spreads electronically for  $V > V_t$ , (disregarding initial transients due to circuit capacitance). Let  $V$  exceed  $V_t$  and as we have shown a single propagated action potential results. Note that it is not recurrent, but we must not overlook the fact that if  $V$  is maintained constant during and after the action potential we will have an electrotonic spread along the line leaving one section of the line from  $x=0$  to  $x=x_t$  above threshold and the section  $x=x_t$  to infinity below threshold but in a state of heightened excitability. From  $x=0$  to  $x=x_t$  by definition  $V(x) > V_t$  and the relay contacts are closed. In the steady state the line is simplified as below.

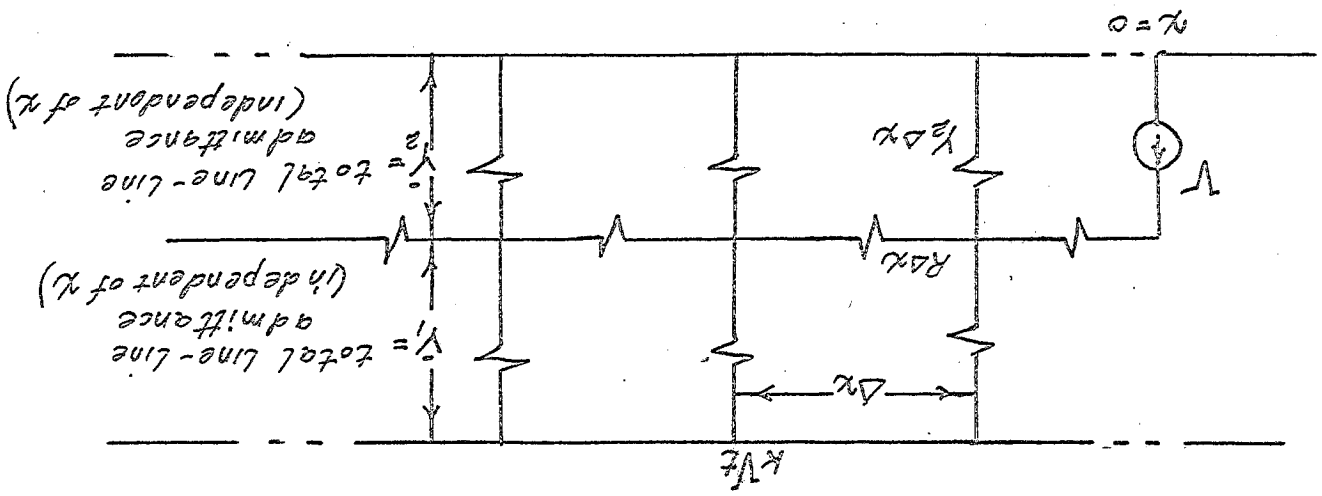


Fig. 16. Resistance network of 'relay' Neuron.

This line is linear for  $V(x) \geq V_t$  so we can justifiably superimpose the effects of the distributed source voltage  $KV_t$  and the input voltage  $V$  for any point  $0 \leq x \leq \infty$

(32)  $V(x) \approx \frac{K}{K+Y} KV_t + V_S(x,0)$  assuming  $\Delta x$  small; i.e. line ~ distributed

(33)  $V = \sqrt{(Y+Y_2)R}$  where  $V_S(x,0) = V_t$  = line voltage at  $x$ , due to  $V$  at  $x=0$

Note that to determine the independent effects of  $V$  and the

energy source  $kV_t$  we let each in turn be zero and add the responses.

When  $kV_t = 0$  we have:

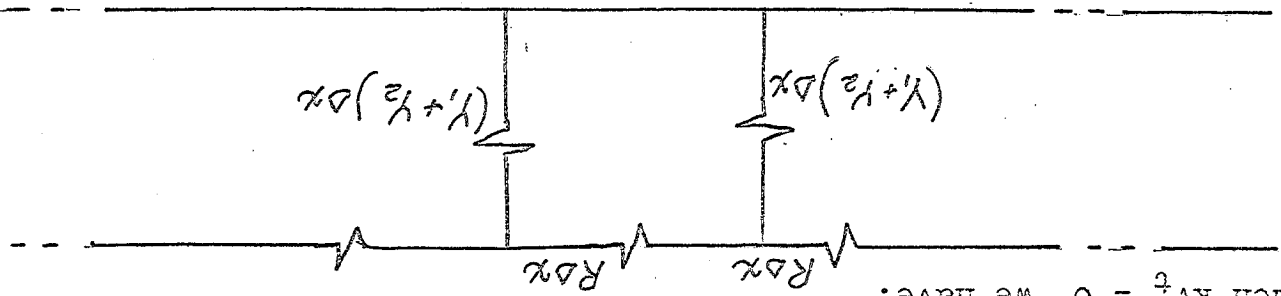


FIG. 17. Reduction of resistive network.

(34)

$$V_s^0(n, 0) = V e^{-Rn}$$

By Superposition

(35)

$$V(n) \approx V e^{-Rn} + \frac{Y}{Y+Y_2} kV_t e^{-Rn}$$

(36)

$$= \left( V - \frac{Y}{Y+Y_2} kV_t \right) e^{-Rn} + \frac{Y}{Y+Y_2} kV_t$$

Note:  $kV_t \frac{Y}{Y+Y_2} > V_t$

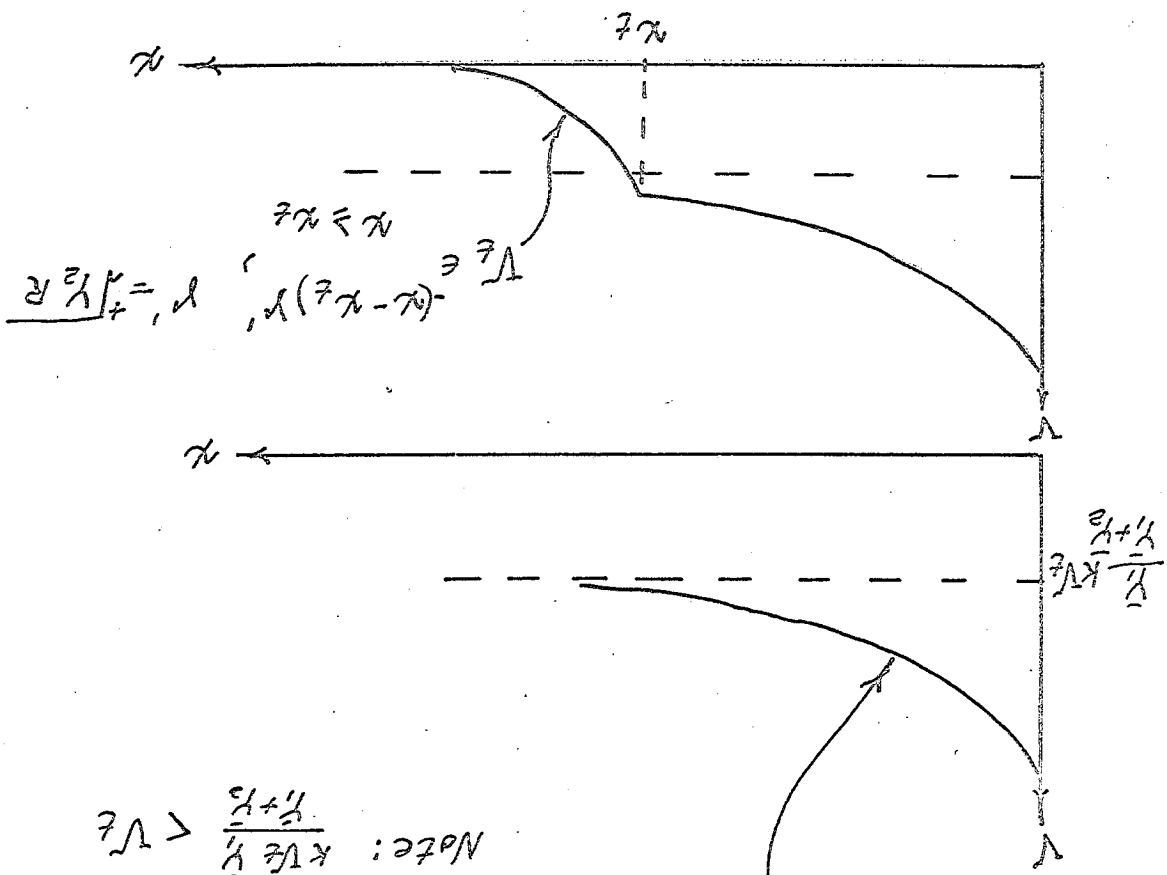


FIG. 18. Voltage along relay neuristor line

The 'resonant' length is then

$$(43) \quad \frac{1}{L} \ln(V - c_2) - c_1 = \omega \tau$$

So that

$$(42) \quad c_2 = \frac{1}{V} k V \tau = c_1$$

Let  $\frac{1}{L} \ln(V - c_1) = c_1$  (independent of  $V$ )

$$(40) \quad \omega \tau = \frac{1}{L} \ln \left( V - \frac{1}{V} k V \tau \right) - \ln \left( V - \frac{1}{V} k V \tau \right)$$

$$(39) \quad \frac{V \tau \left( 1 - \frac{1}{V} k V \tau \right)}{V - \frac{1}{V} k V \tau} = \omega \tau$$

And solving for  $\omega \tau$

$$(38) \quad V(\omega \tau) = V - \frac{1}{V} k V \tau + \frac{1}{V} k V \tau e^{-\omega \tau} + \frac{1}{V} k V \tau e^{-\omega \tau}$$

At  $x = x_1$

$$U(x - x_1) = \text{unit step function at } x = x_1$$

where

$$(37) \quad V(x) = \left[ V - \frac{1}{V} k V \tau \right] e^{-\omega \tau} + \frac{1}{V} k V \tau e^{-\omega \tau} + \frac{1}{V} k V \tau e^{-\omega \tau} + U(x - x_1) e^{-\omega \tau}$$

The above development is applicable for both continuous external and internal reflection types of feedback discussed earlier. As a reminder, in the continuous external feedback case the loop is completed by continuous uni-directional coupling and  $v_1$  and  $v_2$  may be different. Also in the case of repeated internal reflection  $v_1$  must equal  $v_2$ .

Feedback Possibilities

$$f = \left[ \frac{v_1}{L - v_1 t} + \frac{v_2}{L - v_2 t} \right]^{-1} = \left[ L - v_1 t \right]^{-1} \left[ L - v_2 t \right]^{-1} \frac{v_1 + v_2}{v_1 v_2} \quad (45)$$

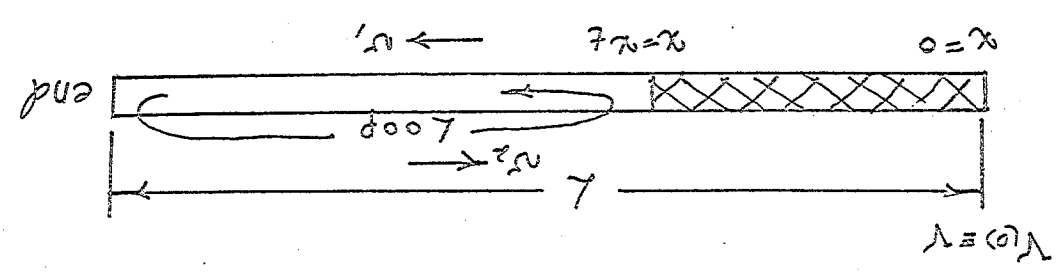
without endpoint delays

time delay  $t_1$  = rebound delay at  $x = x_1$   
time delay  $t_2$  = rebound delay at  $x = L$   
 $f$  = frequency  
 $v_2$  = wave velocity to left via return path  
where  $v_1$  = wave velocity to right

$$f = \left[ \frac{v_1}{L - v_1 t} + \frac{v_2}{L - v_2 t} + \text{delay}_1 + \text{delay}_2 \right]^{-1} \quad (44)$$

at  $x = x_1$  and  $x = L$  the frequency will be  
Assuming that there is reflection exactly

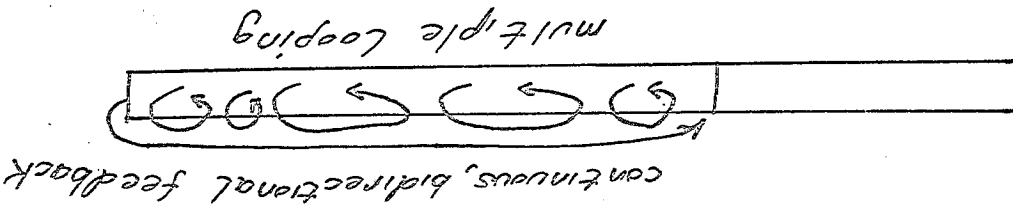
Fig. 19. Internal line resonance.



tained.

this process of oscillation does not cease as long as  $V$  is main-  
as it is below or above threshold respectively. It is assumed that  
spreads electronically with or without attenuation accordingly,  
without gain) For the case of internal reflection the return signal  
feedback path (whether it has active or passive elements, with or  
the return signal level is determined by the characteristics of the  
initiating a new pulse wave to the right. With continuous feedback,  
of sufficient magnitude traveling to the left will reexcite the line  
line is in a condition of heightened excitability. A return signal  
Wave reflection also occurs near the point  $x=x_1$ , because the

Fig. 20. Continuous feedback oscillator.



voltage control will be lost.

be uni-directional or else small multiple loops will appear and  
must begin at this point. In this case the external looping must  
If on the other hand there is continuous feedback, an external loop  
recover (capacitors  $C_1$  in the relay neuristor model are discharged).  
a delay there so that the region immediately behind the wave can  
If it is to be reflected back along the line itself, there must be  
right at velocity  $V_1$  meets a discontinuity at the extreme right.  
delays are possible. The action potential wave traveling to the  
It remains to examine the conditions at the end points where

INHIBITORY INFLUENCES ON RHYTHMICITY

Further investigation of the Neuristor

Up to this point we have been greatly concerned with generation of pulses (action potentials) along the neuron-like line, but for completeness we should also be interested in mechanisms by which such discharges are inhibited. Obviously, removal of the input will prevent rhythmic pulsations from happening, but it is also quite possible to have a super-threshold input and in addition one or more inhibitory inputs which prevent or diminish the tendency for discharges to occur.

To illustrate possible mechanisms for this phenomena let us

investigate two modes of inhibition. Perhaps the most obvious

method of excitatory-inhibitory control is by direct summation of

inputs as shown below.

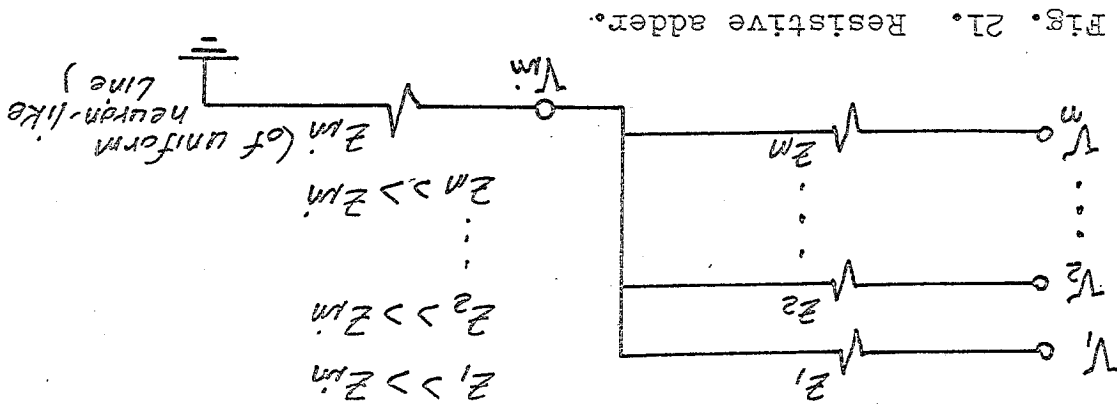


Fig. 21. Resistive adder.

This system requires that the inputs be applied at the same

point on the line and that the inputs greatly exceed the output

$V_{in}$  for strictly additive properties to exist.

There is quite a different method of inhibition which requires

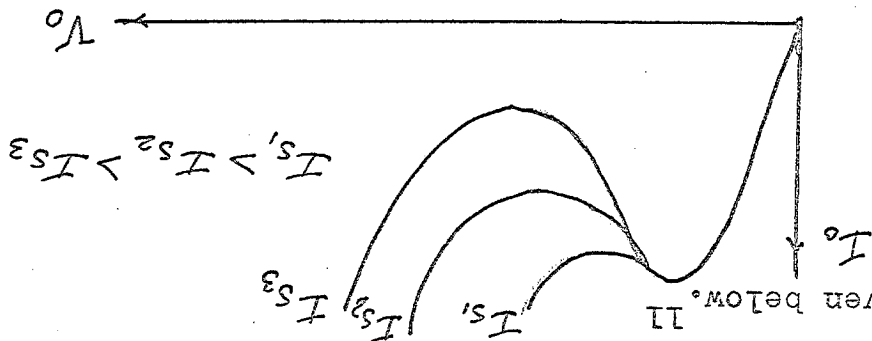
a new point of application. Consider for a moment one version of

the neuristor discrete-element line. There are several possible

line structures that might be chosen but this one is purposely



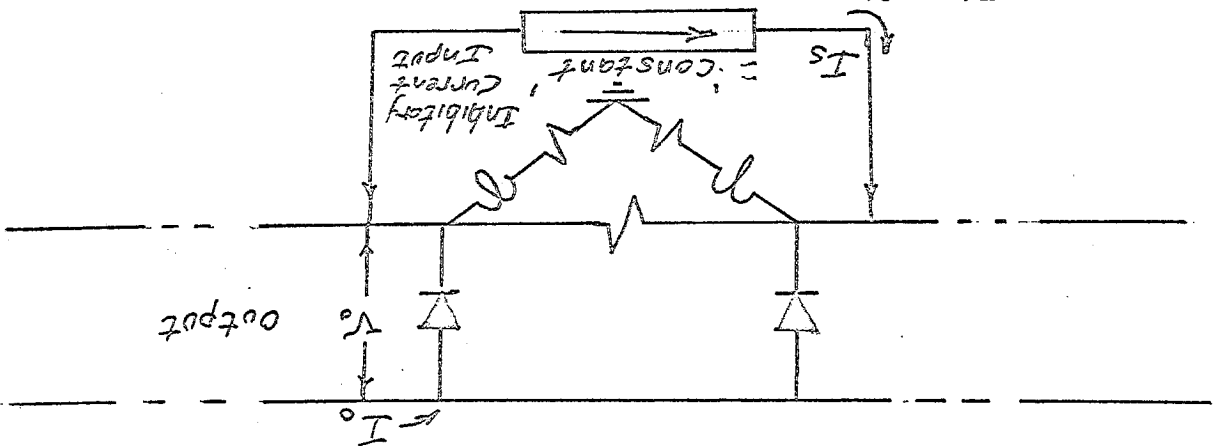
Fig. 25. Tunnel diode characteristics as a function of a control current.



of curves given below. 11  
 Varying  $I_s$  on the output characteristic is illustrated by the family of the externally applied inhibitory input current. The effect of diode characteristics with controlled negative resistance a function

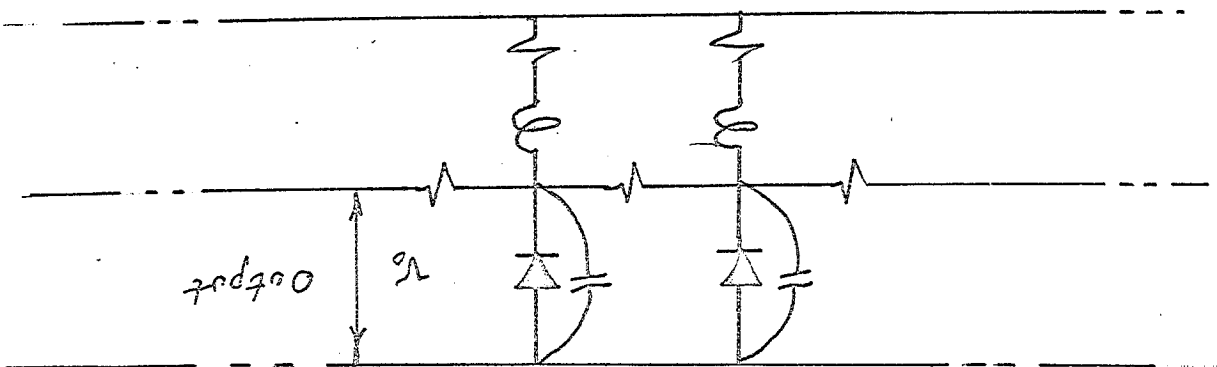
This segment as seen from the output terminals has tunnel

Fig. 24. One form of Inhibitory Input.



of L and C the line may be broken into the section below. selected to illustrate the following. If we ignore the effects

Fig. 23. One version of the neuron.



We note immediately the increase in 'valley' current with increasing inhibitory current  $I_S$ . It should be obvious that this represents a decrease in gain for this segment of the line. If  $I_S$  is sufficiently large an action potential reaching the segment will be attenuated to the level where no further activation of the line is possible and thus it will completely block the signal. On the other hand, if  $V$  is applied at approximately the same point as  $I_S$  the inhibitory effect of  $I_S$  will not be absolute and it merely will raise the apparent threshold. Obviously there are other disturbances which might be introduced on the line to inhibit its transmission. The above are but two conceivable examples.

Shot Noise

Given an input  $V$  and a series of inhibitory currents and/or voltages we assume that the output response of our neuron-like

line is thus uniquely determined. It is generally the rule that

in actual experimental studies of receptor cell response one en-

counters random pulse frequency and amplitude variations which we

have not anticipated for the simple electrical analog. Up to now

such phenomena have not been considered important, but the effects

of randomness in the physiological system are worthy of attention.

It is spontaneous neural activity, actually a form of noise which

stimulates and heightens the receptivity of the system. Also

random statistical fluctuations are an important factor in limiting

the relative accuracy of the pulse frequency coded neural signals.

It is reasonable to suspect that our distributed neuron-like

line will be subject to fluctuations in parameters along its length

as a function of applied voltage and time. We know that all elec-

tronic discrete components possess random variations and are

therefore generators of noise of one form or another. What we do

know of distributed parameter circuitry suggests that this noise

problem is universal. However, relative magnitude must be noted

in these variations.

It is generally accepted that in the neuron longitudinal

resistance/unit length and shunt capacitance/unit length are

'virtually' independent of applied voltage, (as noted earlier). We

would therefore be cautioned to look elsewhere for sources of noise.

The natural parameter to suspect is the nonlinear one. In the

neuron for the logical choice is the parameter with tunnel diode

characteristics.

It is during the tunneling process that shot noise manifests

itself. Shot noise is believed to be the result of carriers de-

flected from the barrier when unlike charges of equal energy levels

are not on opposite sides of the barrier. It is also recognized in

electron tubes as the result of the irregular impact of electrons

on the plate of the tube. Similarly we might expect shot noise dur-

ing ionic conduction across a neural membrane.

At the low frequencies of neuron-like lines (in the audio

range) tunnel diode noise is almost entirely shot noise while ther-

mal noise is present in significant amounts only at high frequencies,

too high above the normal response of such lines. Shot noise is

inversely proportional to frequency (over a wide range) and an

increasing function of voltage.<sup>12</sup>

It is not unreasonable for us to include such phenomena as

shot noise in our neuron-like line model.

<sup>12</sup> S. P. Gentile, *Ibid.*, p. 251.

After familiarizing ourselves with the neuron-like line and some of the road-blocks to straight-forward analysis in our understanding of rhythmic behavior. Powerful 'tools' such as Lyapunov's Second Method were not readily applicable to this problem. As an alternative we searched for a number of network simplifications, based upon quantitative assumptions which we left without justification.

Before actually probing into the neuron-like line, we considered possible mechanisms by which oscillatory behavior could be achieved, thus motivating later discussion. As a first approach, we examined the Hodgkin-Huxley circuit model. Here the shunt conductance was seen to vary with respect to distance away from the point of input application. This variation was used to fabricate a roughly equivalent R-C series configuration. From this model a frequency vs. voltage dependence was established with neuron excitation strength-duration characteristics.

In an entirely different manner we next pursued the notion that transmission line resonance was a possible mode of oscillation. To this end we utilized a simplified discrete element, piece-wise linear approximation of the Neuron transmission line. Our analysis showed that the voltage along such a line delays exponentially with respect to distance, with different asymptotes for segments of the line above and below threshold voltage. Near the transition point the line was on the verge of initiating an action potential. Although the actual stability at this point was

#### SUMMARY

not considered, it was reasoned that a returning pulse of sufficient magnitude would elicit a new action potential. Thus this heightened excitability could furnish the necessary wave regeneration for repetitive behavior. Both internal line reflection and external return signals were included in our study of line resonance. In the later case we introduced the concept of continuous feedback.

At this point we considered negative influences upon rhythmicity and the effect of multiple inputs both excitatory and inhibitory. Pulse blocking was studied for the case of the tunnel diode Neuron. We concluded with a consideration of statistical variation and noise.

It must be realized that almost any simplification that we might desire to make of the complex neuron-like line is likely to mislead us. As previously stated it is only out of necessity that we are compelled to perhaps over-simplify our model. Ultimately it is required that all conclusions reached be investigated experimentally and/or analytically in the most realistic model we have. Thus if we suspect some feedback mechanism, for example, in the animal receptor cell, it might be useful to study the effect of breaking the feedback loop, if this is possible. Unfortunately, nerve studies are difficult or impossible in all but the largest cells such as the giant squid axon due to the problems inherent in manipulating tissue measured in micron dimensions.

With neuristor lines we are confronted with the problem of physical realizability. Here it is required to synthesize nonlinear distributed parameter lines. The synthesis of lumped parameter networks with nonlinear properties, even with the aid of computer facilities, is indeed a formidable task. Transform methods are only useful for piecewise-linear networks so that ultimately we must work exclusively in the time domain.

Then again, assuming that the synthesis problem has been completely resolved theoretically on paper, it then remains to develop the component characteristics and circuitry to carry out the design. Then we may desire to develop a distributed model in the laboratory. Neuristor development awaits technological breakthroughs in distributed circuitry.

It is hoped that this work will encourage others to expand neuron analogs so that new comparisons can be made with their animal counterparts.

#### UNSOLVED PROBLEMS

#### CONCLUSION

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